



Receiver-Based Bayesian PAPR Reduction in OFDM

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High PAPR in OFDM Signals

- ▶ Though OFDM signals are used in many wireless communication standards they suffer from high PAPR.
- ▶ The high PAPR causes either of the following
 - ▷ Nonlinear distortions from the high power amplifier (HPA).
 - ▷ Inefficient use of HPA due to power backoff in order to avoid distortions.

Existing Schemes for PAPR Reduction

- ▶ There are numerous existing transmitter based schemes including
 - ▷ Selected mapping (SLM)
 - ▷ Partial transmit sequence (PTS)
 - ▷ Tone reservation (TR)
- ▶ These schemes are computationally expensive
- ▶ Cause reduction in battery lifetime

Contributions

- ▶ The proposed scheme is
 - ▷ Bayesian and uses the *a priori* knowledge of the signal statistics.
 - ▷ weighted to focus on most probable clipped locations.
 - ▷ phase augmented to acknowledge the anti-phase nature of clipping signal.
 - ▷ able to refine the initial estimates of the signal statistics if exact estimates are not available.

Results

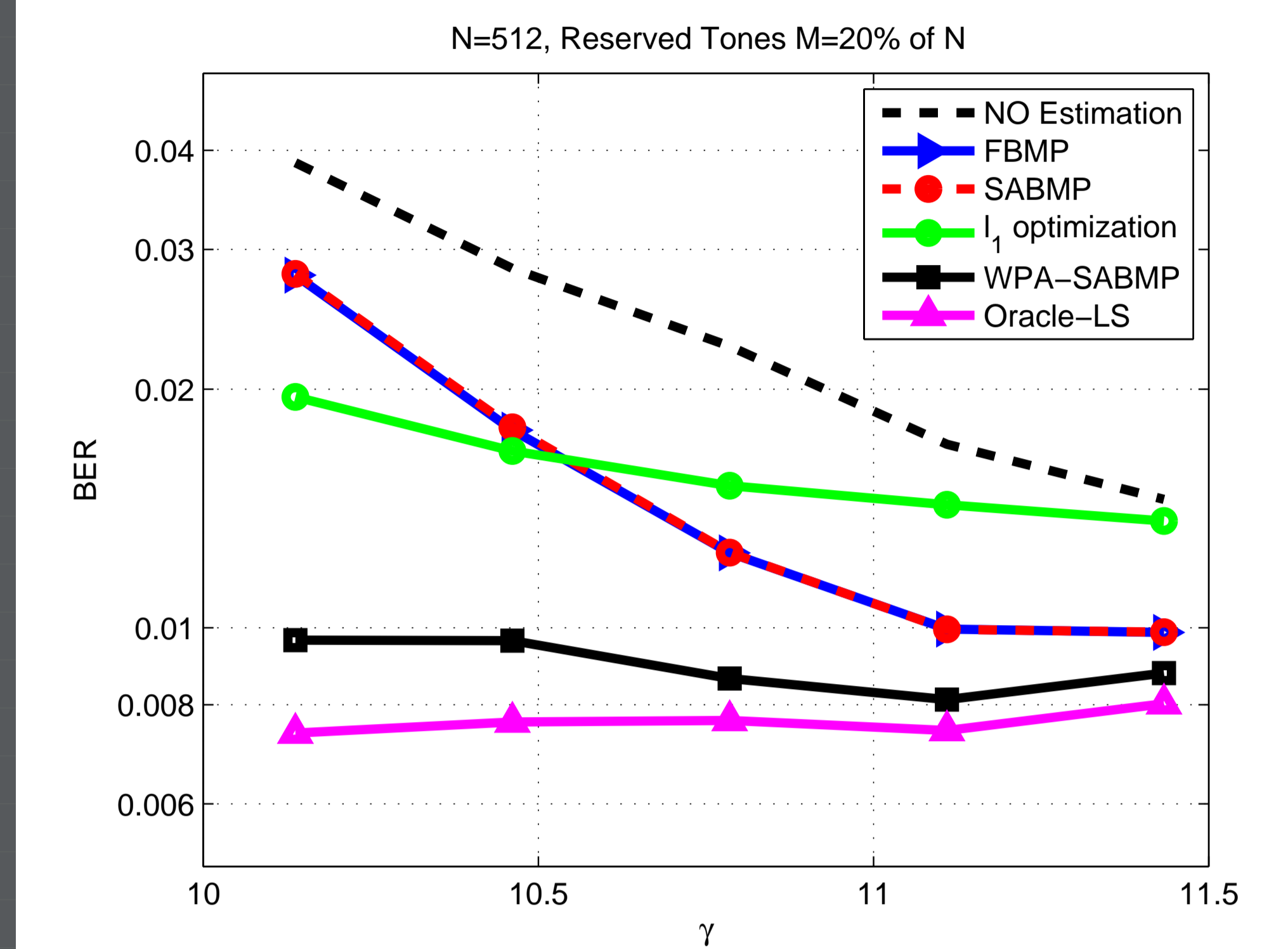


Figure: BER versus γ using exact parameter values

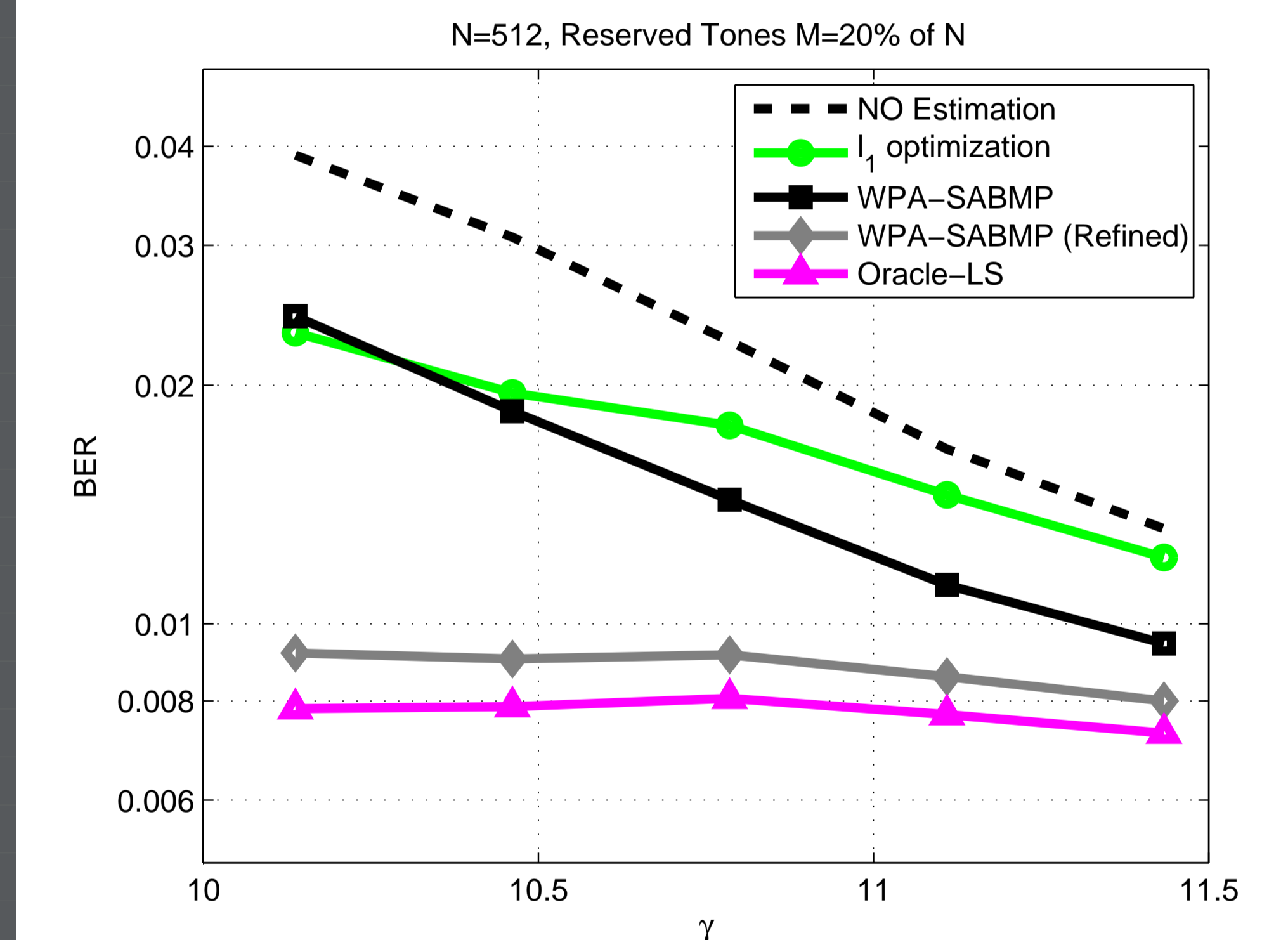


Figure: BER versus γ using rough estimates of the parameters

References

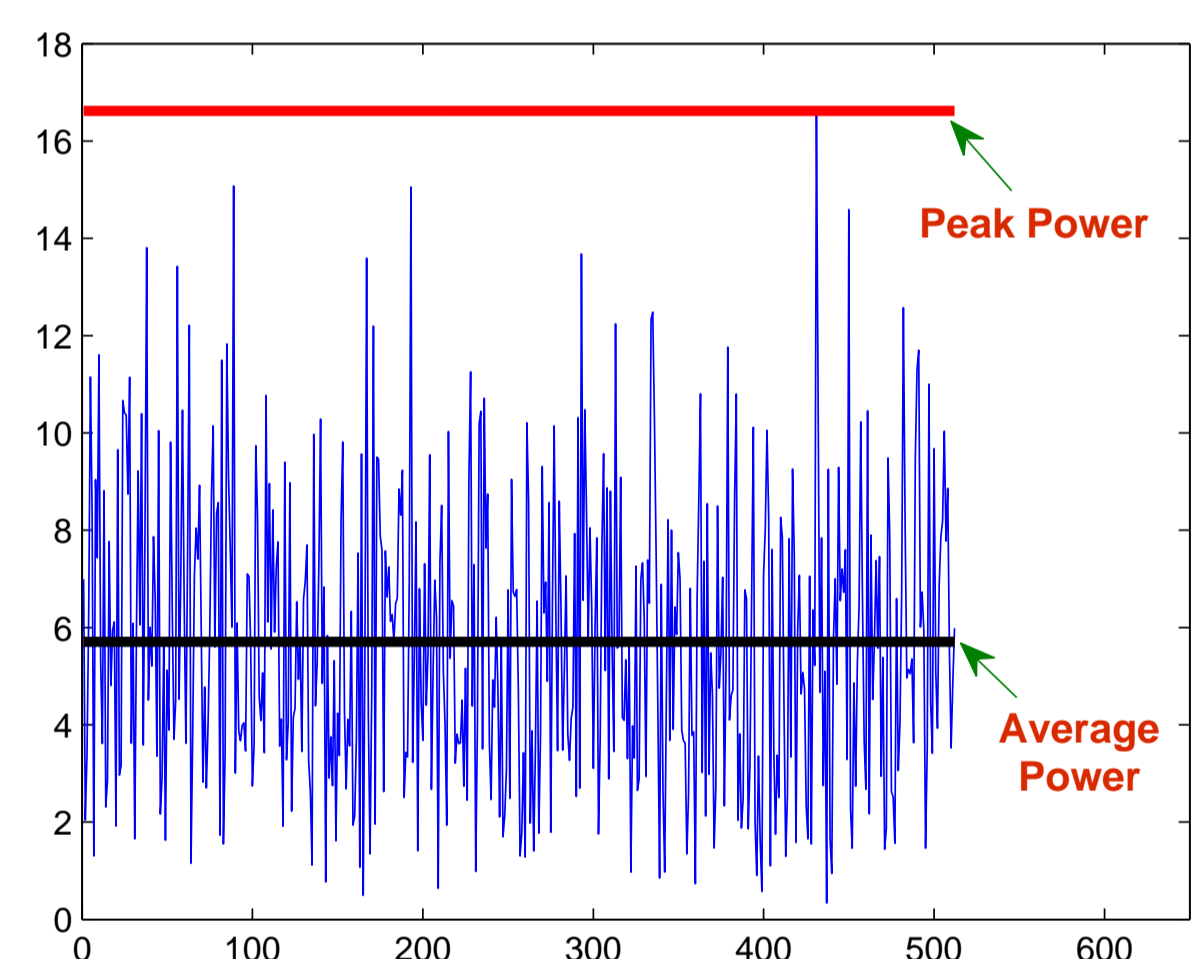
- [1] E. Candes and J. Romberg, "Sparsity and incoherence in compressive sampling," *Inverse problems*, vol. 23, no. 3, p. 969, 2007.
- [2] M. Masood and T. Y. Al-Naffouri, "Sparse Reconstruction using Distribution Agnostic Bayesian Matching Pursuit," *IEEE Trans. Signal Process.*

Clipping at the Transmitter

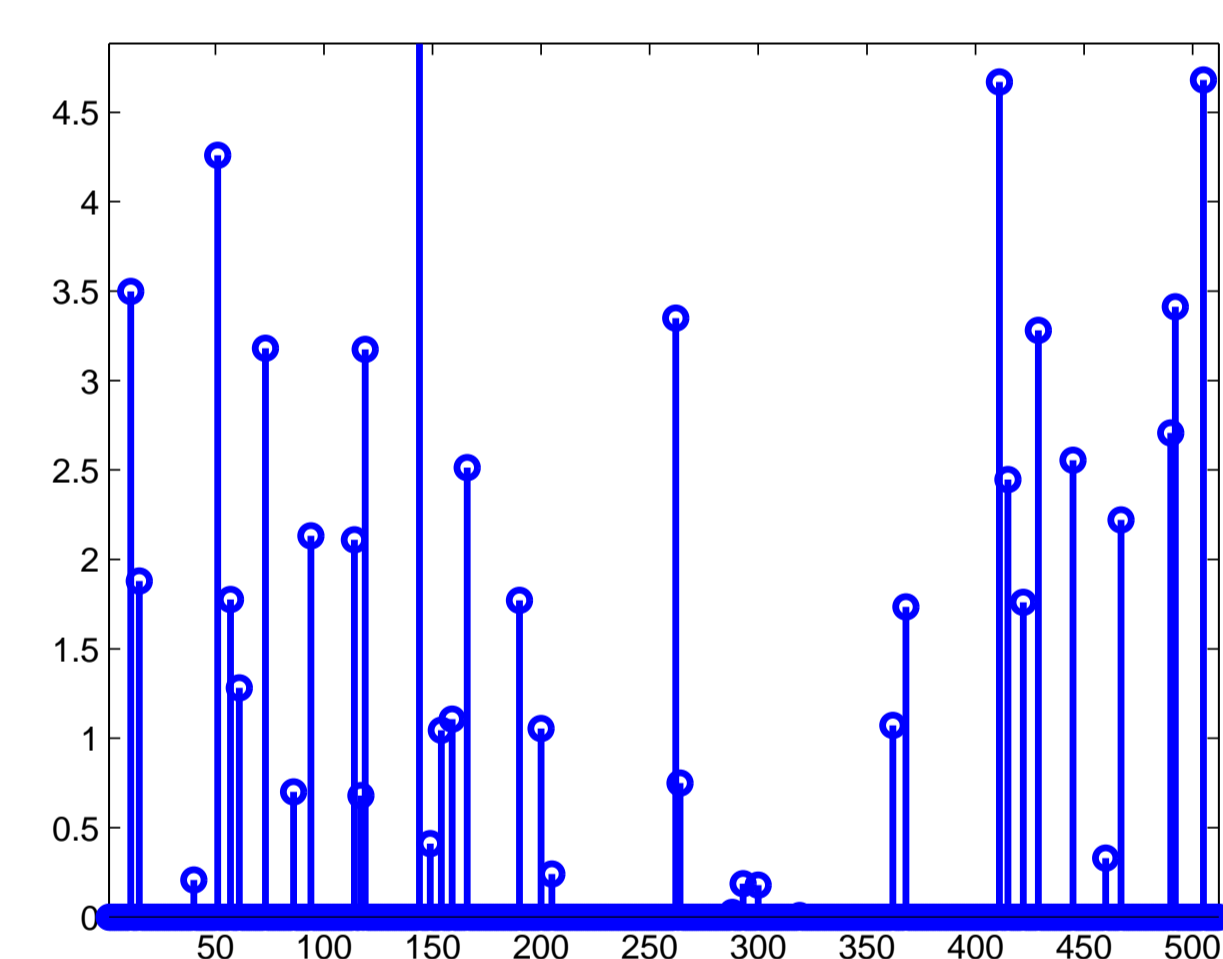
- ▶ A clipping signal c is added to the the time domain high PAPR signal x to get the clipped signal x_p

$$x_p(i) = \begin{cases} \gamma e^{j\theta_{x(i)}} & \text{if } |x(i)| > \gamma \\ x(i) & \text{otherwise} \end{cases}$$

- ▶ The amplitude of the clipped signal x_p is limited to clipping threshold γ .
- ▶ The clipping signal c is sparse in time domain.

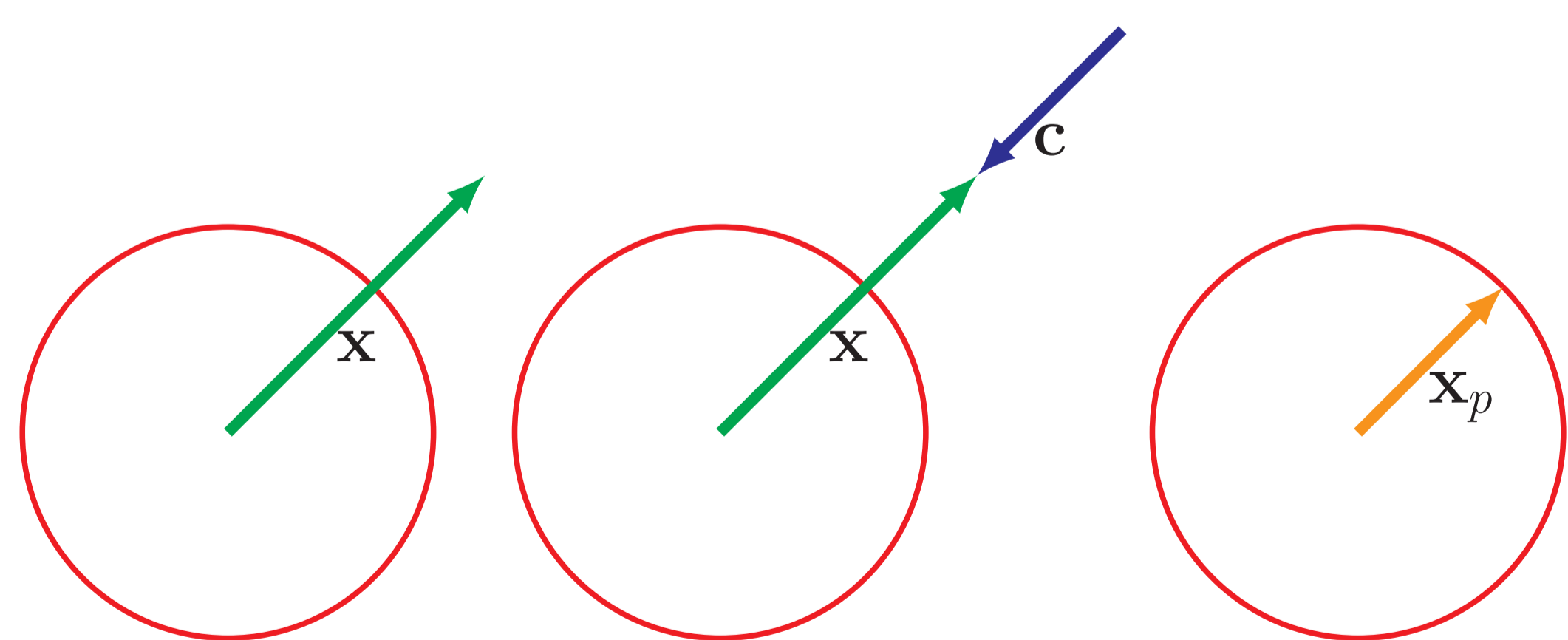


(a) High PAPR OFDM Signal



(b) Sparse Clipping Signal c

- ▶ The clipping signal c only changes the magnitude of the signal x and not the phase (c is anti-phased with x).



- ▶ With intention of recovering the clipping signal at the receiver, $K \ll N$ carriers are reserved for sparse signal recovery.
- ▶ These few subcarriers are enough to estimate the sparse vector c [1].

Reconstruction at the Receiver

- ▶ Only the magnitude of the unknown signal c is to be found.
- ▶ Consider a set of equations

$$\mathbf{y}'_m = \Phi_m \mathbf{c} + \mathbf{z}'_m$$

where $\mathbf{y}'_m \in \mathbb{C}^M$, $\Phi_m \in \mathbb{C}^{M \times N}$ and $\mathbf{z}'_m \in \mathbb{C}^M$.

- ▶ The phase of complex c is merged into Φ_m , making the unknown $c \in \mathbb{R}^N$ (and hence the term phase augmented).
- ▶ Hence the above system can be split as

$$\begin{bmatrix} \text{Re}\{\mathbf{y}'_m\} \\ \text{Im}\{\mathbf{y}'_m\} \end{bmatrix} = \begin{bmatrix} \text{Re}\{\Phi'_m\} \\ \text{Im}\{\Phi'_m\} \end{bmatrix} \mathbf{c} + \begin{bmatrix} \text{Re}\{\mathbf{z}'_m\} \\ \text{Im}\{\mathbf{z}'_m\} \end{bmatrix}$$

$$\mathbf{y} = \Phi \mathbf{c} + \mathbf{z}$$

- ▶ The system of equation given above is solved using a CS scheme for clipping recovery.

Weighted Reconstruction

- ▶ For sparse signal reconstruction, most probable locations of active elements are found.
- ▶ Initially each location has a success probability ρ .
- ▶ The signals elements close to γ have higher probability of being active.
- ▶ We use $\mathbf{w} = \gamma - |\hat{\mathbf{x}}_p|$ as a weighting vector
- ▶ Assign higher probabilities to locations where $w(i)$ is small

$$p(\mathcal{S}) = \prod_i p_i, \text{ for all } i = 1, 2, \dots, N$$

where $p_i = \rho e^{-w(i)}$.

Proposed Reconstruction Scheme

- 1) estimate $\hat{\mathbf{x}}_p = \mathbf{F}^H \mathbf{\Lambda}^{-1} \mathbf{F} \mathbf{y}$
- 2) $\hat{\gamma} = \max(\hat{\mathbf{x}}_p)$.
- 3) $\hat{\sigma}_n^2 = \text{var}(\mathcal{Y})$.
- 4) $\mathbf{w} = \hat{\gamma} - |\hat{\mathbf{x}}_p|$.
- 5) $\hat{\rho}_0 = Q\left(\frac{\hat{\gamma} - \mu}{\sigma}\right)$, an initial estimate, where μ and σ are the mean and standard deviation of $\hat{\mathbf{x}}_p$, respectively.
- 6) $i = 0$, **repeat**
- 7) $p_k = \hat{\rho}_i e^{-w(k)}$, $k = 1, 2, \dots, N$.
- 8) **Compute** $\hat{\mathbf{c}}_{\text{ammse}}$ and $\hat{\rho}_{i+1}$ using the technique discussed in [2]
- 9) **until** $\left(\frac{|\hat{\rho}_i - \hat{\rho}_{i-1}|}{\hat{\rho}_{i-1}} < 0.02\right)$
- 10) $\hat{\mathbf{c}} = \Theta_{\mathbf{c}} |\hat{\mathbf{c}}_{\text{ammse}}|$
- 11) $\hat{\mathbf{x}} = \hat{\mathbf{x}}_p - \hat{\mathbf{c}}$