



# Receiver-Based Bayesian PAPR Reduction in OFDM

Abdullatif Al-Rabah, Mudassir Masood, Anum Ali and Tareq Y. Al-Naffouri



Department of Electrical Engineering, King Abdullah University of Science and Technology (KAUST), Saudi Arabia.  
Department of Electrical Engineering, King Fahd University of Petroleum and Minerals (KFUPM), Saudi Arabia.

## High PAPR in OFDM Signals

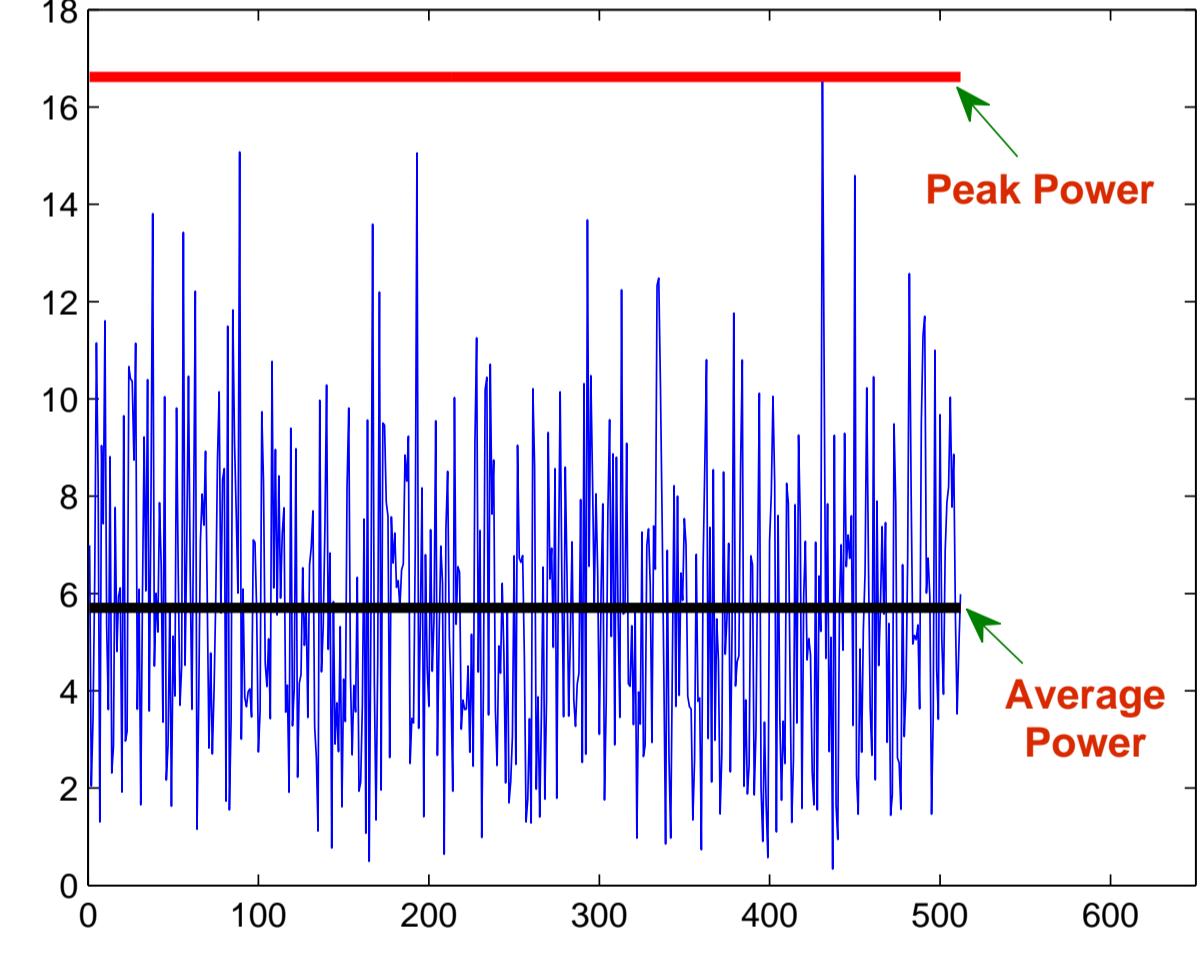
- Though OFDM signals are used in many wireless communication standards they suffer from high PAPR.
- The high PAPR causes either of the following
  - Nonlinear distortions from the high power amplifier (HPA).
  - Inefficient use of HPA due to power backoff in order to avoid distortions.

## Clipping at the Transmitter

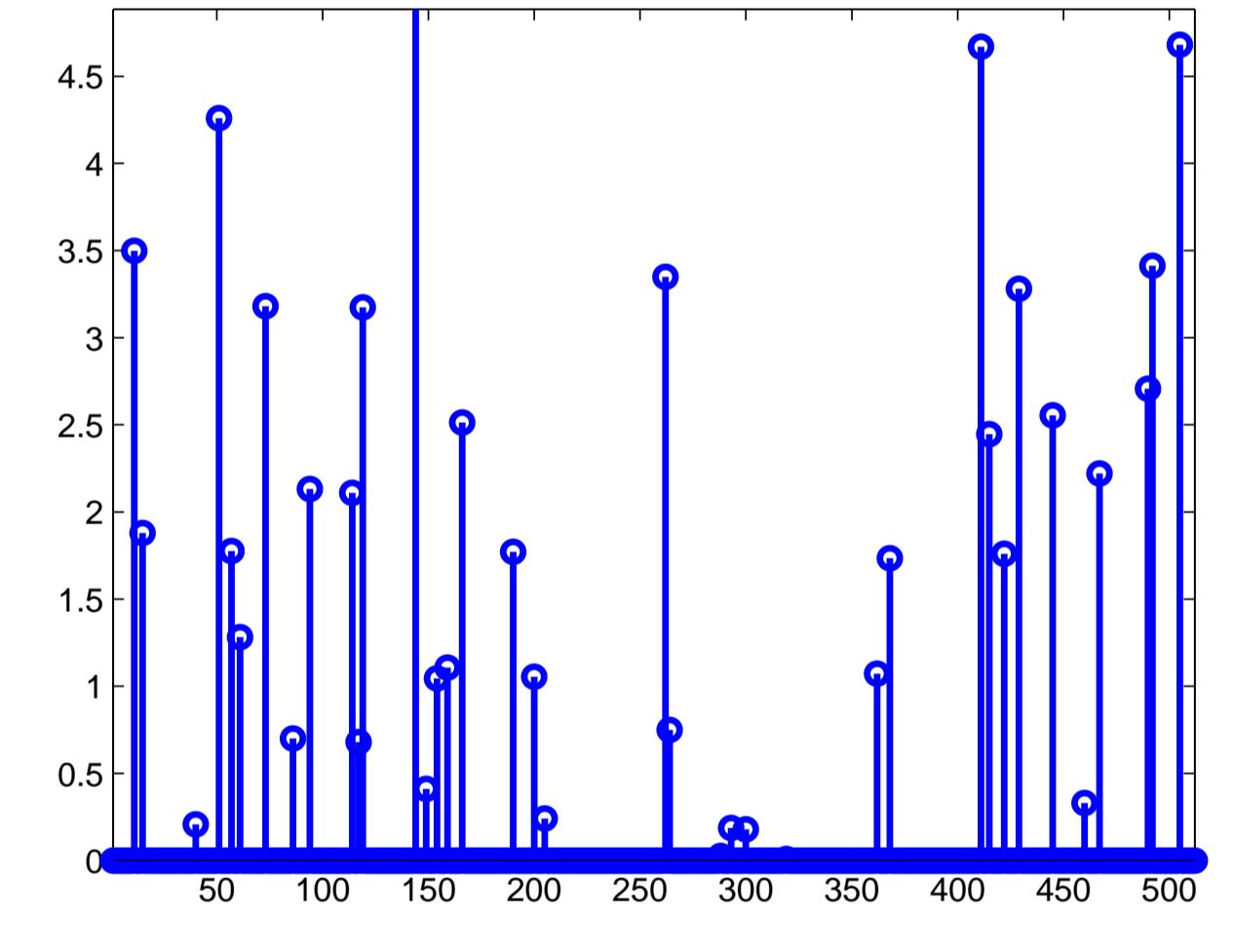
- A clipping signal  $\mathbf{c}$  is added to the time domain high PAPR signal  $\mathbf{x}$  to get the clipped signal  $\mathbf{x}_p$

$$x_p(i) = \begin{cases} \gamma e^{j\theta_{x(i)}} & \text{if } |x(i)| > \gamma \\ x(i) & \text{otherwise} \end{cases}$$

- The amplitude of the clipped signal  $\mathbf{x}_p$  is limited to clipping threshold  $\gamma$ .
- The clipping signal  $\mathbf{c}$  is sparse in time domain.

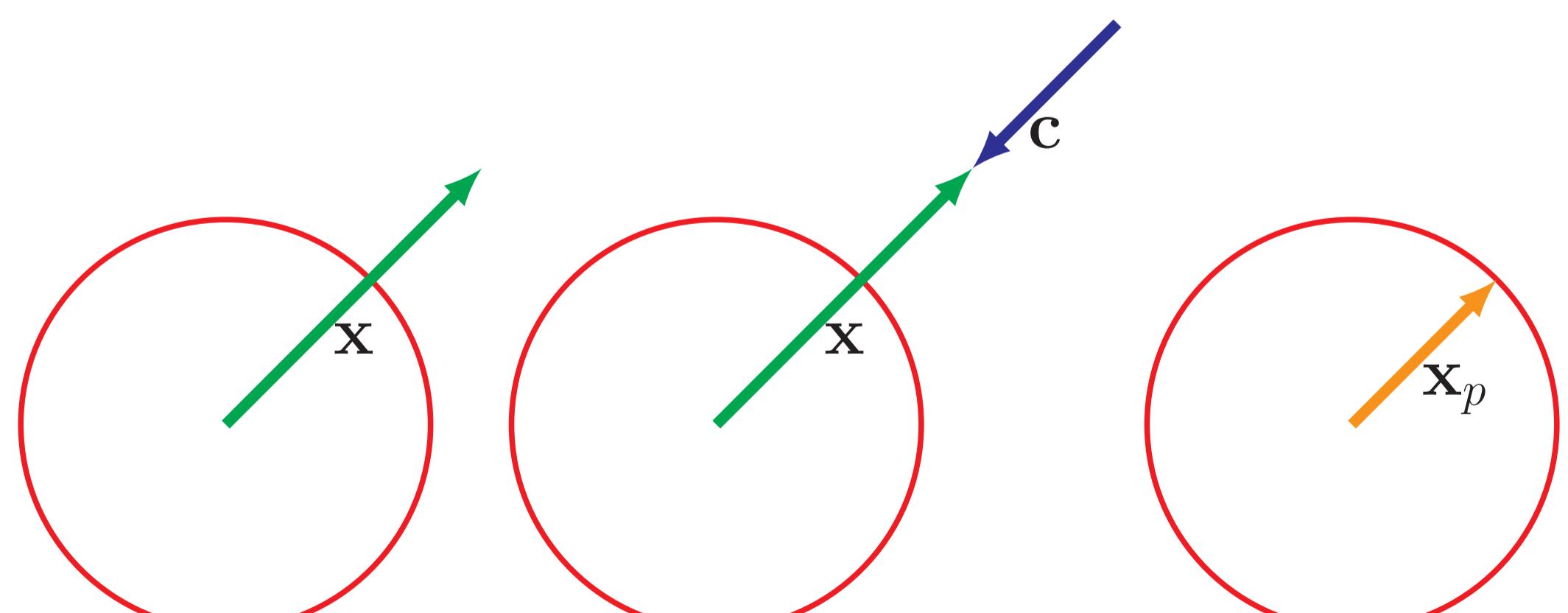


(a) High PAPR OFDM Signal



(b) Sparse Clipping Signal  $\mathbf{c}$

- The clipping signal  $\mathbf{c}$  only changes the magnitude of the signal  $\mathbf{x}$  and not the phase ( $\mathbf{c}$  is anti-phased with  $\mathbf{x}$ ).



- With intention of recovering the clipping signal at the receiver,  $K \ll N$  carriers are reserved for sparse signal recovery.
- These few subcarriers are enough to estimate the sparse vector  $\mathbf{c}$  [1].

## Existing Schemes for PAPR Reduction

- There are numerous existing transmitter based schemes including
  - Selected mapping (SLM)
  - Partial transmit sequence (PTS)
  - Tone reservation (TR)
- These schemes are computationally expensive
- Cause reduction in battery lifetime

## Contributions

- The proposed scheme is
  - Bayesian and uses the *apriori* knowledge of the signal statistics.
  - weighted to focus on most probable clipped locations.
  - phase augmented to acknowledge the anti-phase nature of clipping signal.
  - able to refine the initial estimates of the signal statistics if exact estimates are not available.

## Reconstruction at the Receiver

- Only the magnitude of the unknown signal  $\mathbf{c}$  is to be found.
- Consider a set of equations

$$\mathbf{y}'_m = \Phi_m \mathbf{c} + \mathbf{z}'_m$$

where  $\mathbf{y}'_m \in \mathbb{C}^M$ ,  $\Phi_m \in \mathbb{C}^{M \times N}$  and  $\mathbf{z}'_m \in \mathbb{C}^M$ .

- The phase of complex  $\mathbf{c}$  is merged into  $\Phi_m$ , making the unknown  $\mathbf{c} \in \mathbb{R}^N$  (and hence the term phase augmented).
- Hence the above system can be split as

$$\begin{bmatrix} \text{Re}\{\mathbf{y}'_m\} \\ \text{Im}\{\mathbf{y}'_m\} \end{bmatrix} = \begin{bmatrix} \text{Re}\{\Phi'_m\} \\ \text{Im}\{\Phi'_m\} \end{bmatrix} \mathbf{c} + \begin{bmatrix} \text{Re}\{\mathbf{z}'_m\} \\ \text{Im}\{\mathbf{z}'_m\} \end{bmatrix}$$

$$\mathbf{y} = \bar{\Phi} \mathbf{c} + \bar{\mathbf{z}}$$

- The system of equation given above is solved using a CS scheme for clipping recovery.

## Weighted Reconstruction

- For sparse signal reconstruction, most probable locations of active elements are found.
- Initially each location has a success probability  $\rho$ .
- The signals elements close to  $\gamma$  have higher probability of being active.
- We use  $\mathbf{w} = \gamma - |\hat{\mathbf{x}}_p|$  as a weighting vector
- Assign higher probabilities to locations where  $w(i)$  is small

$$p(\mathcal{S}) = \prod_i^N p_i, \text{ for all } i = 1, 2, \dots, N$$

where  $p_i = \rho e^{-w(i)}$ .

## Proposed Reconstruction Scheme

- estimate  $\hat{\mathbf{x}}_p = \mathbf{F}^H \Lambda^{-1} \mathbf{F} \mathbf{y}$
- $\hat{\gamma} = \max(\hat{\mathbf{x}}_p)$ .
- $\hat{\sigma}_n^2 = \text{var}(\hat{\mathbf{y}})$ .
- $\mathbf{w} = \hat{\gamma} - |\hat{\mathbf{x}}_p|$ .
- $\hat{\rho}_o = Q\left(\frac{\hat{\gamma} - \mu}{\sigma}\right)$ , an initial estimate, where  $\mu$  and  $\sigma$  are the mean and standard deviation of  $\hat{\mathbf{x}}_p$ , respectively.
- $i = 0$ , repeat
- $p_k = \hat{\rho}_i e^{-w(k)}$ ,  $k = 1, 2, \dots, N$ .
- Compute  $\hat{\mathbf{c}}_{\text{ammse}}$  and  $\hat{\rho}_{i+1}$  using the technique discussed in [2]
- until  $\left(\frac{|\hat{\rho}_i - \hat{\rho}_{i-1}|}{\hat{\rho}_{i-1}}\right) < 0.02$
- $\hat{\mathbf{c}} = \Theta_c |\hat{\mathbf{c}}_{\text{ammse}}|$
- $\hat{\mathbf{x}} = \hat{\mathbf{x}}_p - \hat{\mathbf{c}}$

## Results

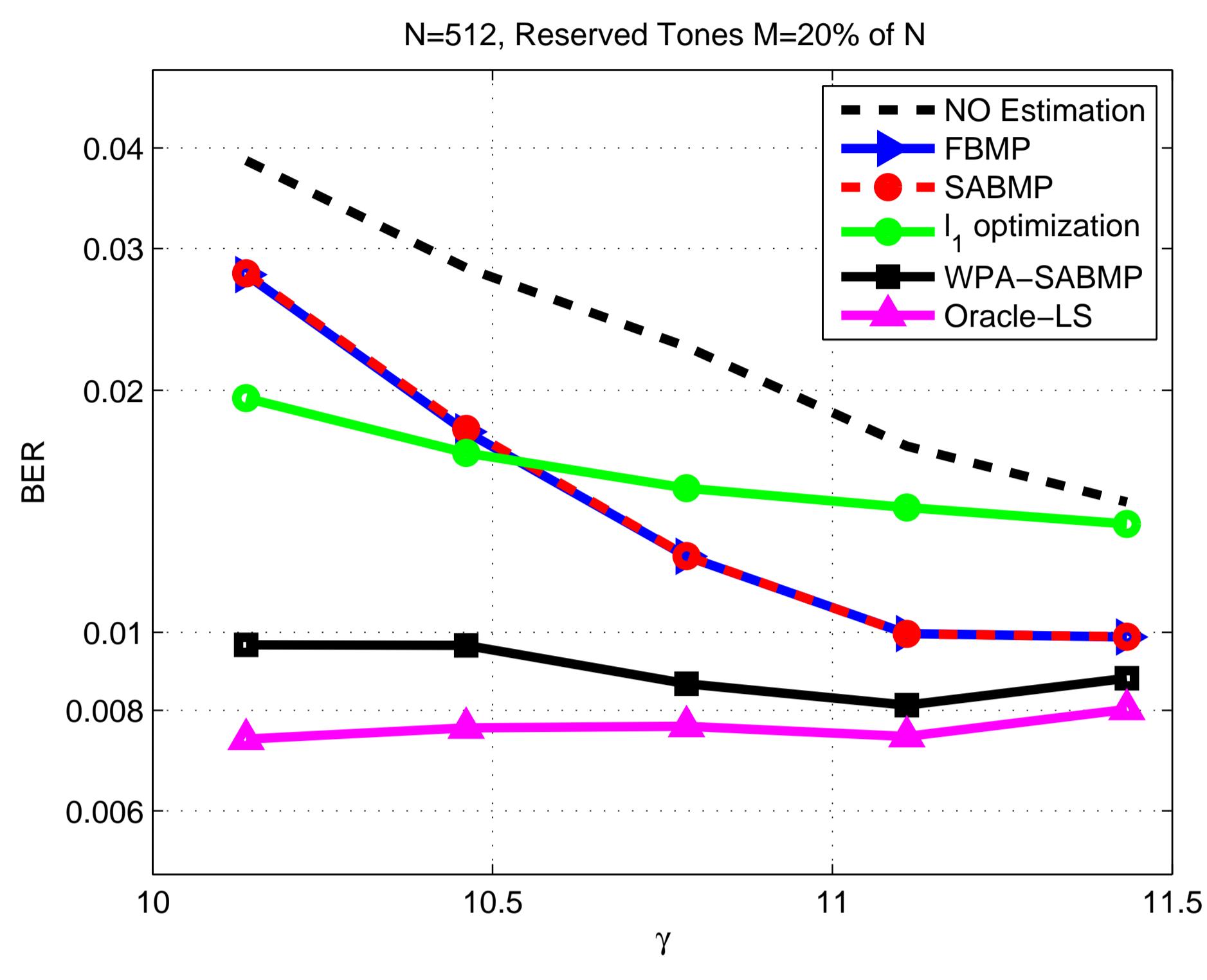


Figure: BER versus  $\gamma$  using exact parameter values

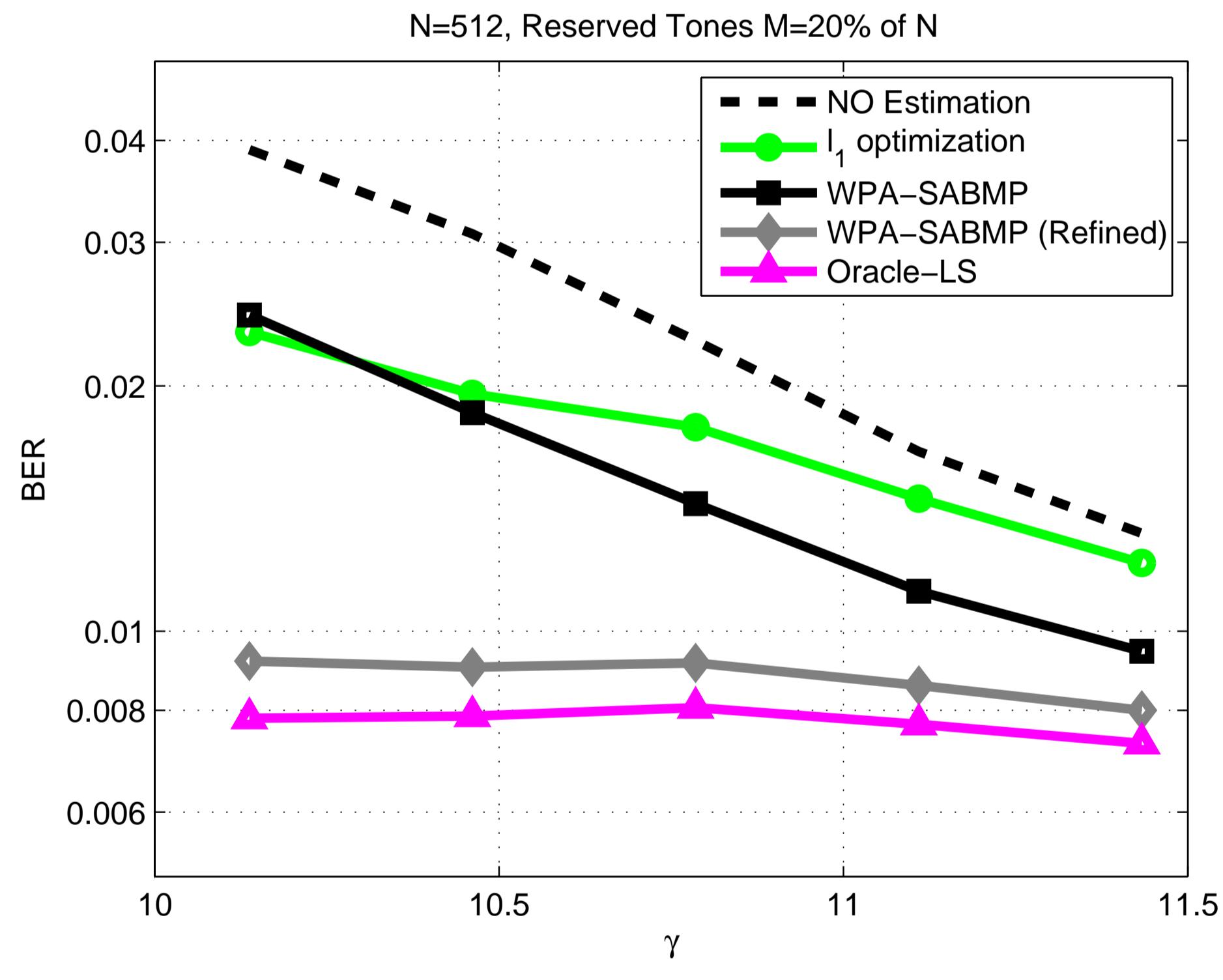


Figure: BER versus  $\gamma$  using rough estimates of the parameters

## References

- E. Candes and J. Romberg, "Sparsity and incoherence in compressive sampling," *Inverse problems*, vol. 23, no. 3, p. 969, 2007.
- M. Masood and T. Y. Al-Naffouri, "Sparse Reconstruction using Distribution Agnostic Bayesian Matching Pursuit," *IEEE Trans. Signal Process.*